

## Узкополосные сигналы

$$s(t) = A_s(t) \cos \omega_0 t - B_s(t) \sin \omega_0 t \quad (1)$$

$$\begin{aligned} s(t) \cos \omega_0 t &= A_s(t) \cos^2 \omega_0 t - B_s(t) \sin \omega_0 t \cos \omega_0 t = \\ &= 0,5A_s(t) + 0,5A_s(t) \cos 2\omega_0 t - 0,5B_s(t) \sin 2\omega_0 t \end{aligned} \quad (2)$$

$$\begin{aligned} s(t) \sin \omega_0 t &= A_s(t) \cos \omega_0 t \sin \omega_0 t - B_s(t) \sin^2 \omega_0 t = \\ &= -0,5B_s(t) + 0,5A_s(t) \sin 2\omega_0 t + 0,5B_s(t) \cos 2\omega_0 t \end{aligned} \quad (3)$$

## Комплексная огибающая (низкочастотный эквивалент)

$$\dot{s}_{\text{НЧ}}(t) = A_s(t) + jB_s(t) \quad (4)$$

$$s(t) = \operatorname{Re}(\dot{s}_{\text{НЧ}}(t)e^{j\omega_0 t}) = A_s(t) \cos \omega_0 t - B_s(t) \sin \omega_0 t \quad (5)$$

$$\dot{s}_{\text{НЧ}}(t) = U_s(t)e^{j\varphi_s(t)} \quad (6)$$

$$U_s(t) = |\dot{s}_{\text{НЧ}}(t)| = \sqrt{A_s^2(t) + B_s^2(t)} \quad (7)$$

$$\varphi_s(t) = \arctg \frac{B_s(t)}{A_s(t)} \quad (8)$$

$$A_s(t) = U_s(t) \cos \varphi_s(t) \quad (9)$$

$$B_s(t) = U_s(t) \sin \varphi_s(t) \quad (10)$$

$$s(t) = U_s(t) \cos(\omega_0 t + \varphi_s(t)) = U_s(t) \cos \psi_s(t) \quad (11)$$

$$\psi_s(t) = \omega_0 t + \varphi_s(t) \quad (12)$$

$$\omega_s(t) = \frac{d\psi_s(t)}{dt} = \omega_0 + \frac{d\varphi_s(t)}{dt} \quad (13)$$

$$\begin{aligned}
\dot{S}(\omega) &= \int_{-\infty}^{\infty} \operatorname{Re} \left( \dot{s}_{\text{HQ}}(t) e^{j\omega_0 t} \right) e^{-j\omega t} dt = \\
&= \frac{1}{2} \dot{S}_{\text{HQ}}(\omega - \omega_0) + \frac{1}{2} \overset{*}{S}_{\text{HQ}}(-\omega - \omega_0)
\end{aligned} \tag{14}$$

## Аналитический сигнал

$$s(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} \dot{S}(\omega) e^{j\omega t} d\omega = \\ = \frac{1}{2\pi} \int_{-\infty}^0 \dot{S}(\omega) e^{j\omega t} d\omega + \frac{1}{2\pi} \int_0^{\infty} \dot{S}(\omega) e^{j\omega t} d\omega \quad (15)$$

$$\dot{s}_a(t) = \frac{1}{\pi} \int_0^{\infty} \dot{S}(\omega) e^{j\omega t} d\omega \quad (16)$$

$$s(t) = 1/2 \left( \dot{s}_a(t) + \overset{*}{s}_a(t) \right) = \operatorname{Re} \dot{s}_a(t) \quad (17)$$

$$\tilde{s}(t) = \operatorname{Im} \dot{s}_a(t) \quad (18)$$

$$\dot{s}_a(t) = s(t) + j\tilde{s}(t) \quad (19)$$

## Преобразование Гильберта

$$\dot{S}_a(\omega) = \begin{cases} 0, & \omega < 0 \\ 2\dot{S}(\omega), & \omega \geq 0 \end{cases} \quad (20)$$

$$\dot{S}_a(\omega) = \dot{S}(\omega) + j\tilde{S}(\omega) \quad (21)$$

$$\tilde{S}(\omega) = -j \operatorname{sign}(\omega) \dot{S}(\omega) = \begin{cases} j\dot{S}(\omega), & \omega < 0 \\ -j\dot{S}(\omega), & \omega \geq 0 \end{cases} \quad (22)$$

$$\tilde{s}(t) = s(t) * \frac{1}{\pi t} = \frac{1}{\pi} \int_{-\infty}^{\infty} \frac{s(\tau)}{t - \tau} d\tau \quad (23)$$

$$\dot{S}(\omega) = j \operatorname{sign}(\omega) \tilde{S}(\omega) \quad (24)$$

$$s(t) = -\tilde{s}(t) * \frac{1}{\pi t} = \frac{1}{\pi} \int_{-\infty}^{\infty} \frac{\tilde{s}(\tau)}{\tau - t} d\tau \quad (25)$$

## Свойства преобразования Гильберта

$$H[a_1 s_1(t) + a_2 s_2(t)] = a_1 H[s_1(t)] + a_2 H[s_2(t)]; \quad (26)$$

$$H[\alpha] = 0, \quad \alpha = \text{const}; \quad (27)$$

$$H[\cos \omega t] = \text{sign}(\omega) \sin \omega t \quad (28)$$

$$H[\sin \omega t] = -\text{sign}(\omega) \cos \omega t \quad (29)$$

Если  $s(t) = A_s(t) \cos \omega_0 t - B_s(t) \sin \omega_0 t$ , то

$$\tilde{s}(t) = H[s(t)] = B_s(t) \cos \omega_0 t + A_s(t) \sin \omega_0 t \quad (30)$$

$$U_s(t) = |\dot{s}_a(t)| = \sqrt{s^2(t) + \tilde{s}^2(t)} \quad (31)$$

$$\dot{s}_{H^q}(t) = \dot{s}_a(t)e^{-j\omega_0 t} \quad (32)$$